

An INTRODUCTION to SWITCHING ADAPTIVE CONTROL

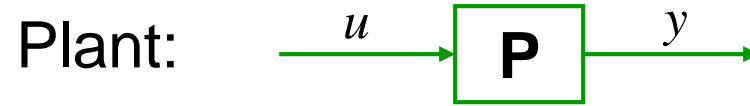
Daniel Liberzon



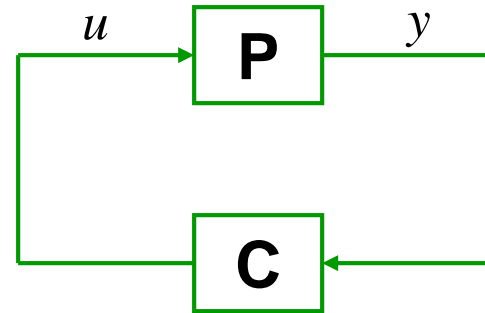
Coordinated Science Laboratory and
Dept. of Electrical & Computer Eng.,
Univ. of Illinois at Urbana-Champaign

Based on joint work with J.P. Hespanha (UCSB) and A.S. Morse (Yale)

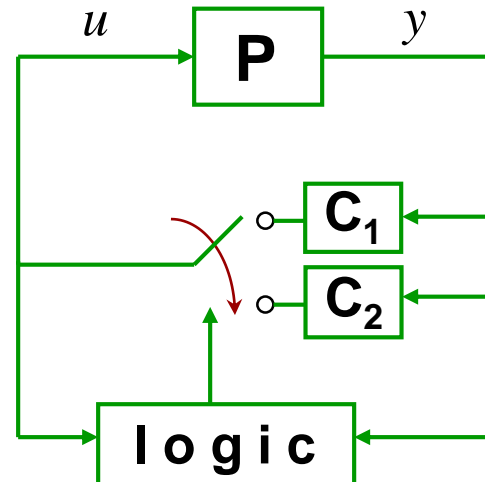
SWITCHING CONTROL



Classical continuous feedback paradigm:



But **logical decisions** are often necessary:



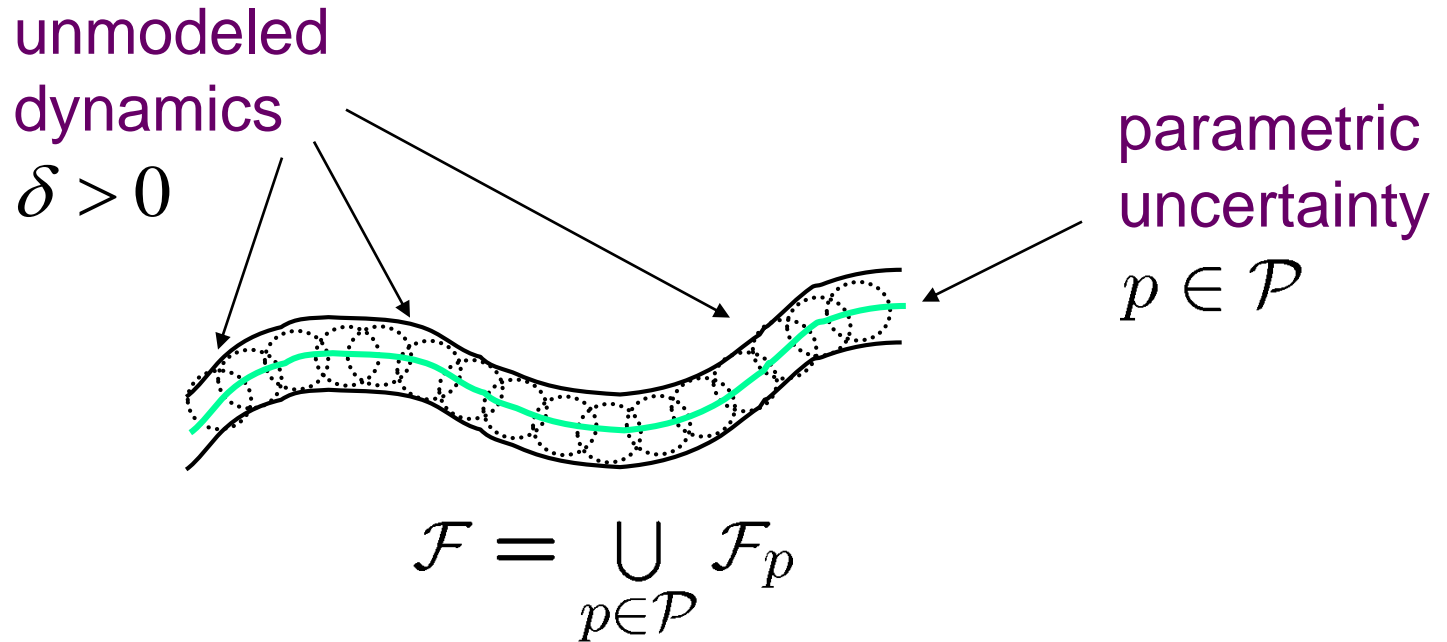
REASONS for SWITCHING

- Nature of the control problem
- Sensor or actuator limitations
- Large modeling uncertainty
- Combinations of the above

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MODELING UNCERTAINTY



Also, **noise** n and **disturbance** d

Adaptive control (continuous tuning)
vs. **supervisory control** (switching)

EXAMPLE

Scalar system:

$$\dot{y} = y^2 + p^* u$$

$$\mathcal{P} = [-10, -0.1] \cup [0.1, 10]$$

$p^* \in \mathcal{P}$, otherwise unknown

(purely parametric uncertainty)

$$u = -\frac{1}{p^*}(y^2 + y) \Rightarrow \dot{y} = -y \quad \text{stable } \checkmark$$

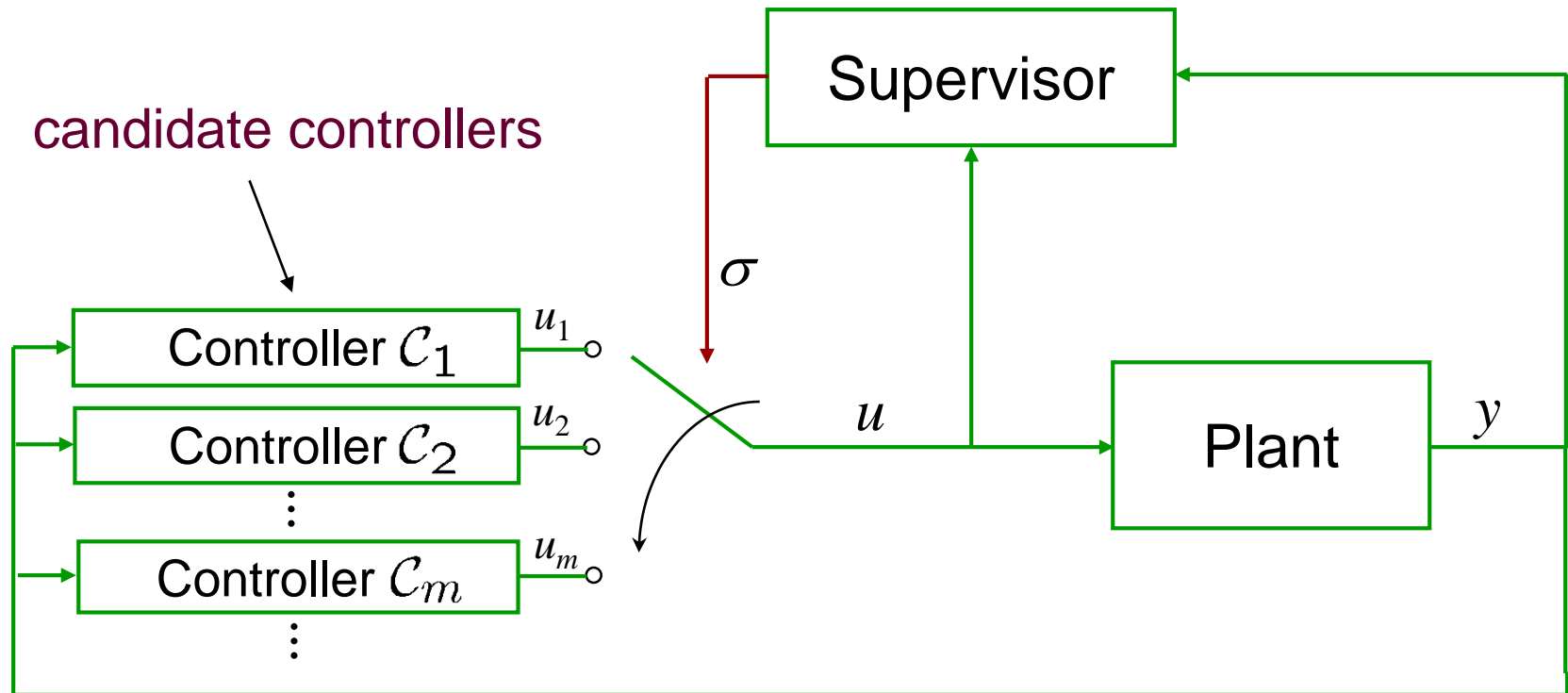
not implementable

Controller family: $u_q = -\frac{1}{q}(y^2 + y)$, $q \in \mathcal{P}$

Could also take u_q , $q \in \mathcal{Q} = \{-1, 1\}$

controller index set

SUPERVISORY CONTROL ARCHITECTURE



σ – switching signal, takes values in \mathcal{Q}

C_σ – switching controller

TYPES of SUPERVISION

- Prescheduled (prerouted)
- Performance-based (direct)
- Estimator-based (indirect)

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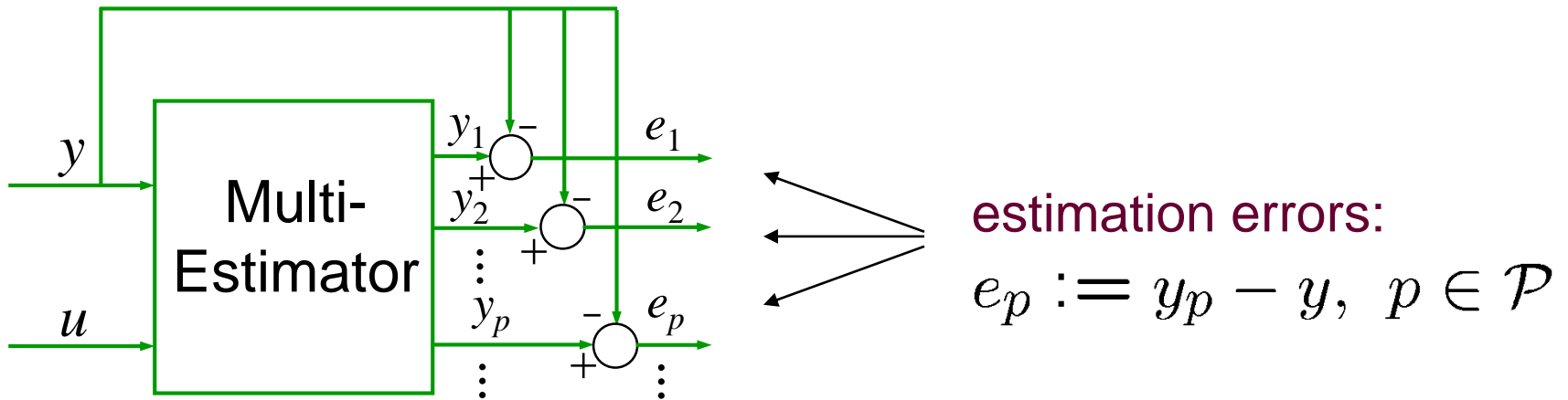
OUTLINE

- Basic components of supervisor
- Design objectives and general analysis
- Achieving the design objectives (highlights)

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SUPERVISOR



Want e_{p^*} to be small

Then e_p small indicates $p = p^*$ likely

EXAMPLE

$$\dot{y} = y^2 + p^* u$$

Multi-estimator:

$$\dot{y}_p = -(y_p - y) + y^2 + pu, \quad p \in \mathcal{P}$$

$$e_p = y_p - y, \quad p \in \mathcal{P}$$



$$\dot{e}_{p^*} = -e_{p^*} \Rightarrow e_{p^*} \rightarrow 0 \text{ exp fast } \forall u$$

EXAMPLE

$$\dot{y} = y^2 + p^*u - d$$

↑ disturbance

Multi-estimator:

$$\dot{y}_p = -(y_p - y) + y^2 + pu, \quad p \in \mathcal{P}$$

$$e_p = y_p - y, \quad p \in \mathcal{P}$$



$$\dot{e}_{p^*} = -e_{p^*} + d \Rightarrow e_{p^*} \rightarrow d \text{ exp fast } \forall u$$

STATE SHARING

$$\dot{y}_p = -(y_p - y) + y^2 + pu, \quad p \in \mathcal{P}$$

Bad! Not implementable if \mathcal{P} is infinite

The system

$$\dot{z}_1 = -z_1 + y + y^2$$

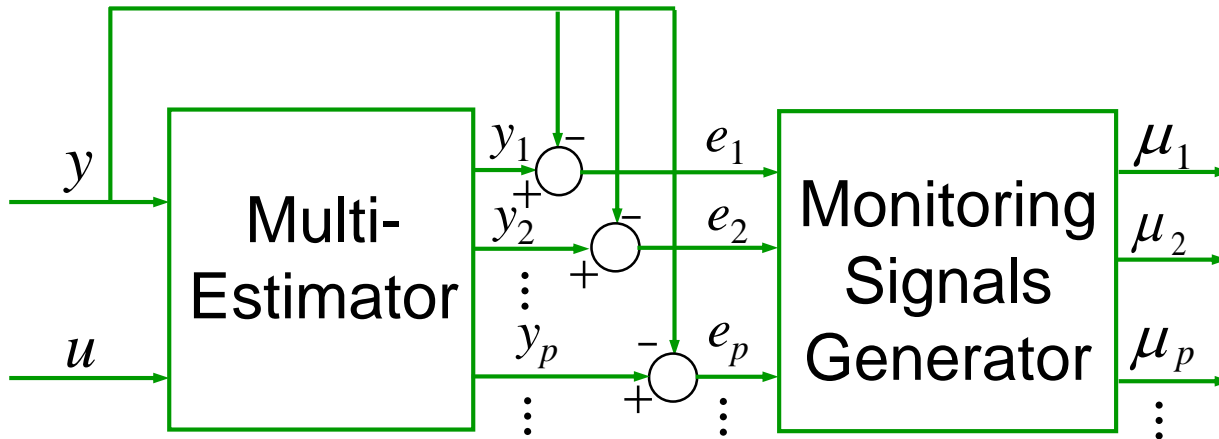
$$\dot{z}_2 = -z_2 + u$$

$$y_p = z_1 + pz_2, \quad p \in \mathcal{P}$$

produces the same signals

$$\dot{y}_p = \dot{z}_1 + p\dot{z}_2 = \underbrace{-z_1 + y + y^2}_{-y_p} + \underbrace{-pz_2 + pu}_{+y_p} = -y_p + y + y^2 + pu$$

SUPERVISOR



Examples:

$$\mu_p(t) = \int_0^t |e_p(\tau)|^2 d\tau \Leftrightarrow \dot{\mu}_p = |e_p|^2, \mu_p(0) = 0$$

$$\mu_p(t) = \int_0^t e^{-\lambda(t-\tau)} |e_p(\tau)|^2 d\tau \Leftrightarrow \dot{\mu}_p = -\lambda\mu_p + |e_p|^2, \mu_p(0) = 0$$

EXAMPLE

Multi-estimator:

$$\dot{z}_1 = -z_1 + y + y^2$$

$$\dot{z}_2 = -z_2 + u$$

$$y_p = z_1 + pz_2, \quad p \in \mathcal{P}$$

$$\dot{\mu}_p = e_p^2 \quad - \text{ can use state sharing}$$

$$e_p^2 = (z_1 + pz_2 - y)^2 = (z_1 - y)^2 + \underbrace{2pz_2(z_1 - y)} + \underbrace{p^2z_2^2}$$

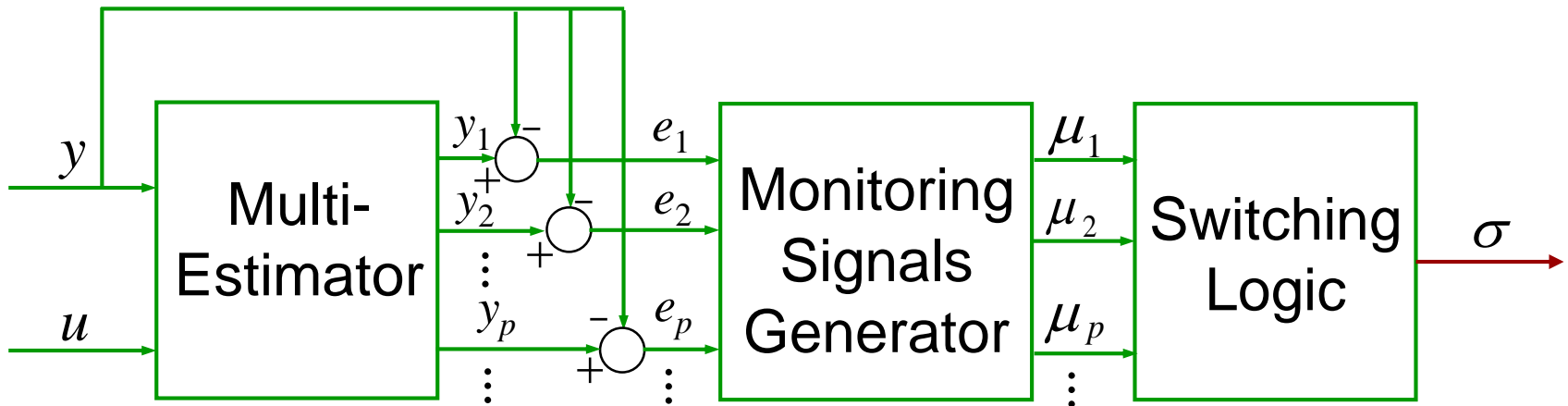
$$\dot{\eta}_1 = (z_1 - y)^2$$

$$\dot{\eta}_2 = 2z_2(z_1 - y)$$

$$\dot{\eta}_3 = z_2^2$$

$$\mu_p = \eta_1 + p\eta_2 + p^2\eta_3, \quad p \in \mathcal{P}$$

SUPERVISOR



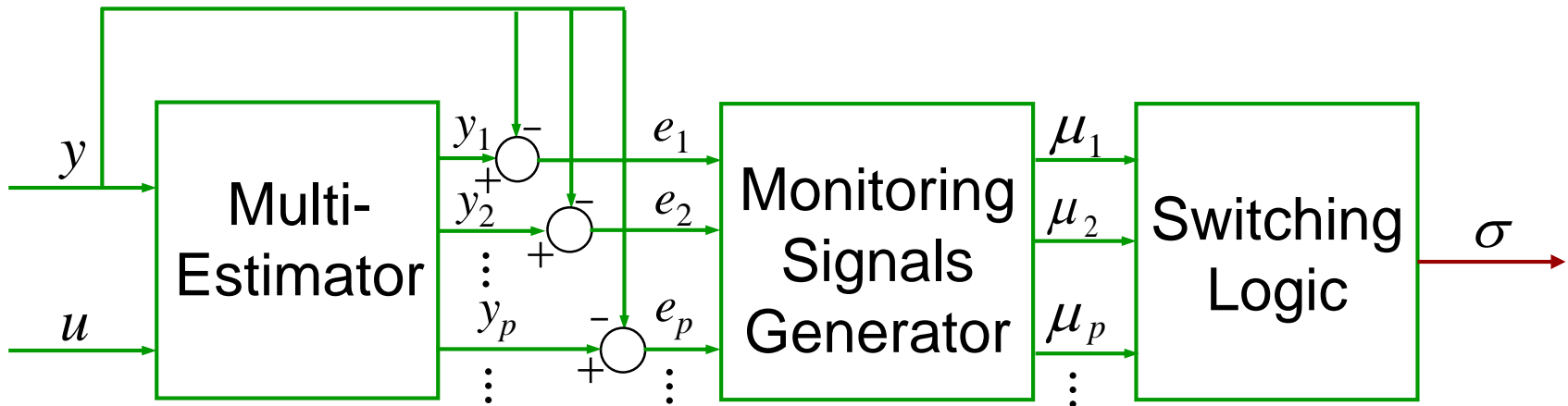
Basic idea: $\sigma(t) = \arg \min_{p \in \mathcal{P}} \mu_p(t)$

Justification? Plant $\in \mathcal{F} = \bigcup_{p \in \mathcal{P}} \mathcal{F}_p$, controllers: $\mathcal{C}_p, p \in \mathcal{P}$

μ_p small $\Rightarrow e_p$ small \Rightarrow plant likely in $\mathcal{F}_p \Rightarrow \mathcal{C}_p$ gives stable closed-loop system

(“certainty equivalence”)

SUPERVISOR



Basic idea: $\sigma(t) = \arg \min_{p \in \mathcal{P}} \mu_p(t)$

Justification? Plant $\in \mathcal{F} = \bigcup_{p \in \mathcal{P}} \mathcal{F}_p$, controllers: $\mathcal{C}_p, p \in \mathcal{P}$

μ_p small $\Rightarrow e_p$ small ~~\Rightarrow~~ plant likely in $\mathcal{F}_p \Rightarrow \mathcal{C}_p$ gives stable closed-loop system
 only know converse!

Need: e_p small $\Rightarrow \mathcal{C}_p$ gives stable closed-loop system

This is **detectability** w.r.t. e_p

DETECTABILITY

Linear case:

$$\dot{x} = A_q x \quad \leftarrow \text{plant in closed loop with } C_q$$

$$e_q = C_q x \quad \leftarrow \text{view as output}$$

Want this system to be **detectable**



$$e_q \rightarrow 0 \Rightarrow x \rightarrow 0$$



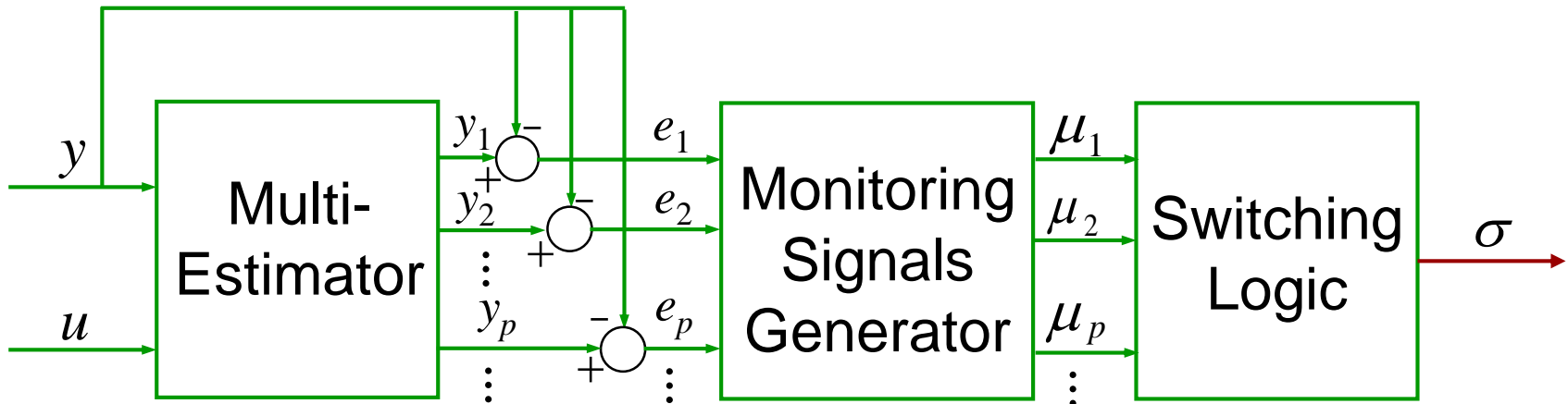
“output injection”
matrix

$$L_q: A_q + L_q C_q \text{ is Hurwitz}$$

$$\dot{\underline{x}} = \underbrace{(A_q + L_q C_q)}_{\text{asympt. stable}} \underline{x} + L_q e_q$$

asympt. stable

SUPERVISOR



We know: e_{p^*} is small

Switching logic (roughly): $\sigma(t) = \arg \min_{p \in \mathcal{P}} \mu_p(t)$

This (hopefully) guarantees that e_σ is small

Need: e_σ small \Rightarrow stable closed-loop switched system

This is **switched detectability**

DETECTABILITY under SWITCHING

Switched system: $\dot{x} = A_\sigma x$ ← plant in closed loop with C_σ

$e_\sigma = C_\sigma x$ ← view as output

Want this system to be **detectable**: $e_\sigma \rightarrow 0 \Rightarrow x \rightarrow 0$

Assumed detectable for each frozen value of σ

Output injection:

$$\dot{x} = \underbrace{(A_\sigma - L_\sigma C_\sigma)}_{\text{need this to be asympt. stable}} x + L_\sigma e_\sigma$$

need this to be asympt. stable

Thus σ needs to be “**non-destabilizing**”:

- switching stops in finite time
- slow switching (on the average)

SUMMARY of BASIC PROPERTIES

Multi-estimator:

1. At least one estimation error (e_{p^*}) is small
 - $e_{p^*} \rightarrow 0 \forall u$ when $n = 0, d = 0, \delta = 0$
 - e_{p^*} is bounded for bounded n & d

Candidate controllers:

2. For each C_q , closed-loop system is detectable w.r.t. e_q

Switching logic:

- ③ e_σ is bounded in terms of the smallest e_p
- ④ Switched closed-loop system is detectable w.r.t. e_σ provided this is true for every frozen value of σ

conflicting: for 3, want to switch to $\arg \min_p \mu_p(t)$
for 4, want to switch slowly or stop

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Switching logic:

3. e_σ is bounded in terms of the smallest e_p
4. Switched closed-loop system is detectable w.r.t. e_σ provided this is true for every frozen value of σ

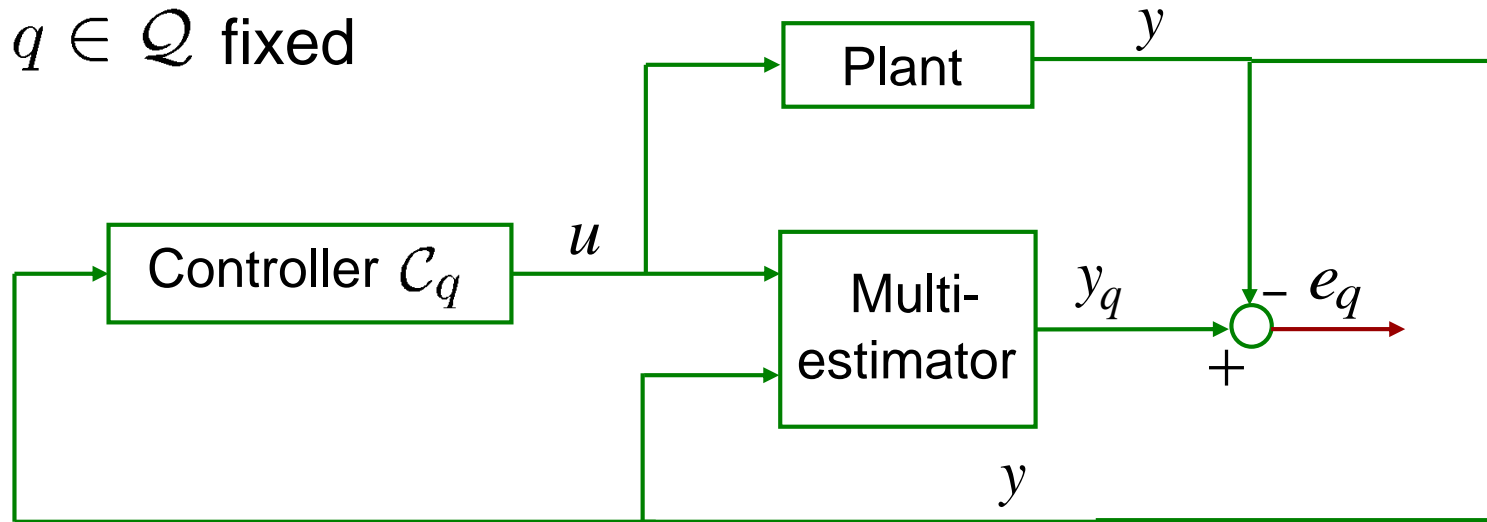
Analysis: $\left. \begin{array}{l} 1 + 3 \Rightarrow e_\sigma \text{ is small} \\ 2 + 4 \Rightarrow \text{detectability w.r.t. } e_\sigma \end{array} \right\} \Rightarrow \text{state is small } \checkmark$

OUTLINE

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- Design objectives and general analysis
- **Achieving the design objectives (highlights)**

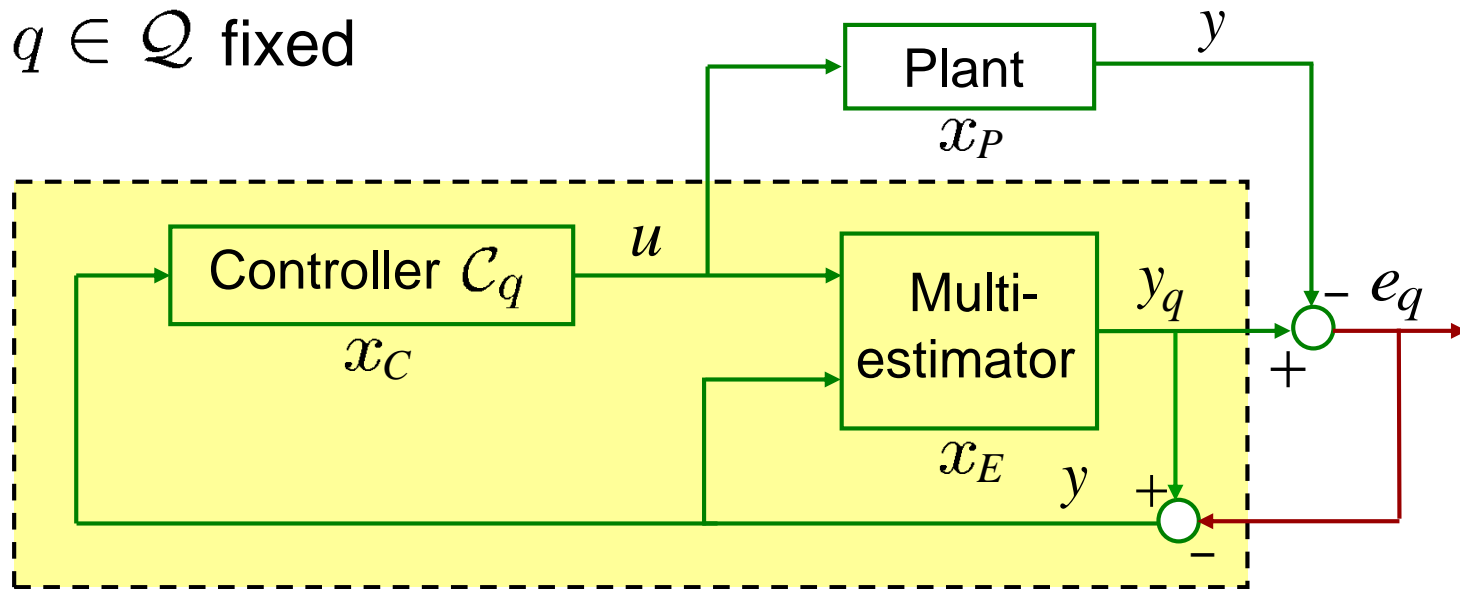
CANDIDATE CONTROLLERS

$q \in \mathcal{Q}$ fixed



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Linear: overall system is detectable w.r.t. e_q if

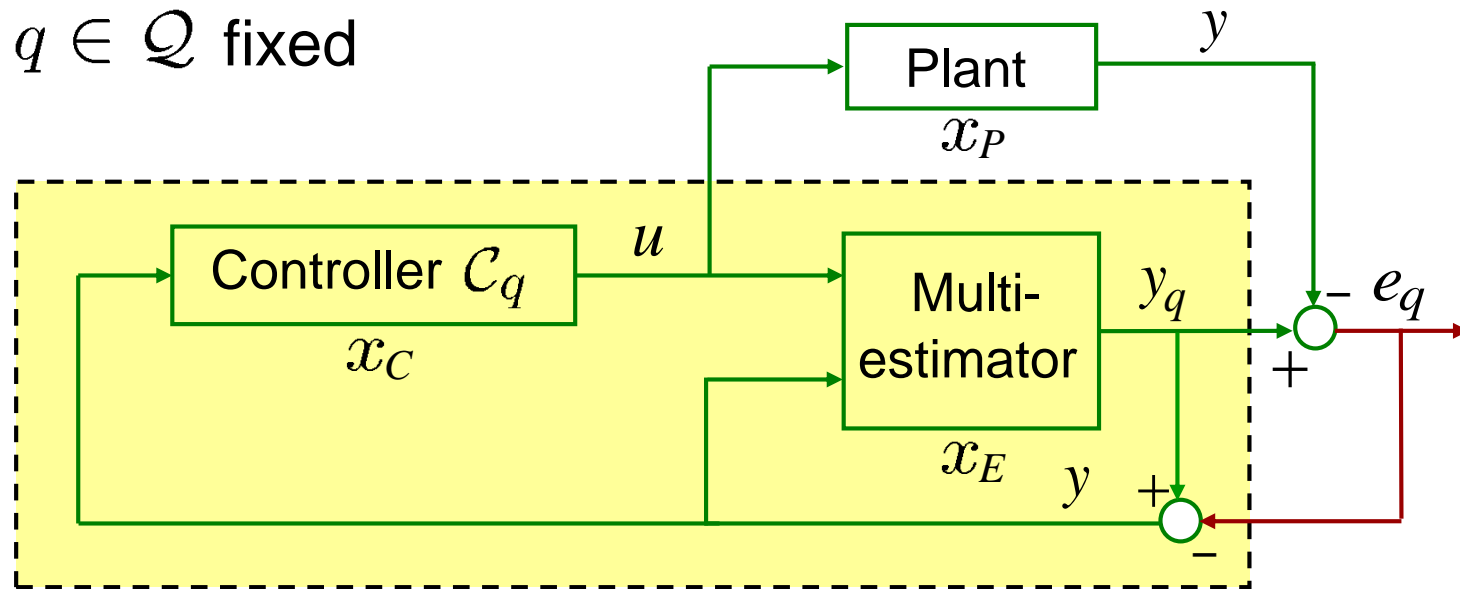
- i. system inside the box is stable
- ii. plant is detectable

Need to show: $e_q \rightarrow 0 \Rightarrow x_P, x_C, x_E \rightarrow 0$

$$e_q \rightarrow 0 \underset{\text{i}}{\Rightarrow} x_C, x_E \rightarrow 0 \Rightarrow u, y_q \rightarrow 0 \Rightarrow y = y_q - e_q \rightarrow 0 \underset{\text{ii}}{\Rightarrow} x_P \rightarrow 0$$

CANDIDATE CONTROLLERS

$q \in \mathcal{Q}$ fixed



Linear: overall system is detectable w.r.t. e_q if

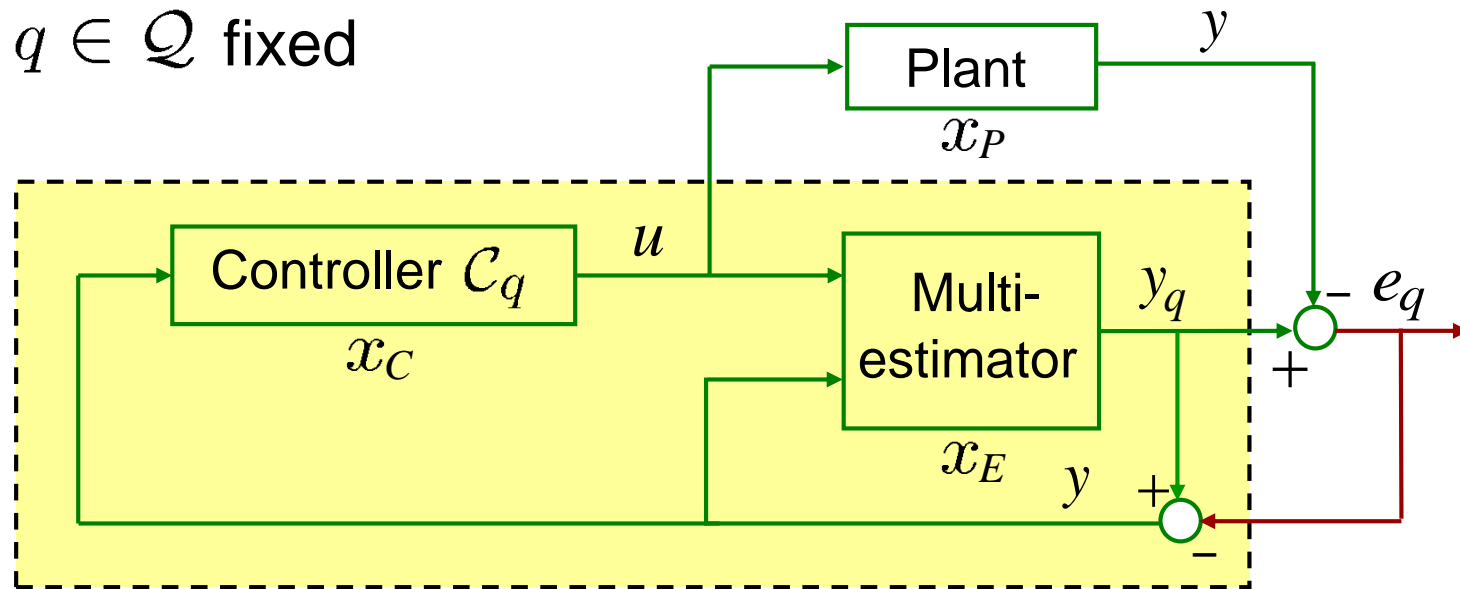
- i. system inside the box is stable
- ii. plant is detectable

Nonlinear: same result holds if stability and detectability are interpreted in the ISS/OSS sense:

$$|x(t)| \leq \beta(|x(0)|, t) + \gamma \left(\| \overset{\text{external signal}}{v} \|_{[0,t]} \right)$$

CANDIDATE CONTROLLERS

$q \in \mathcal{Q}$ fixed



Linear: overall system is detectable w.r.t. e_q if

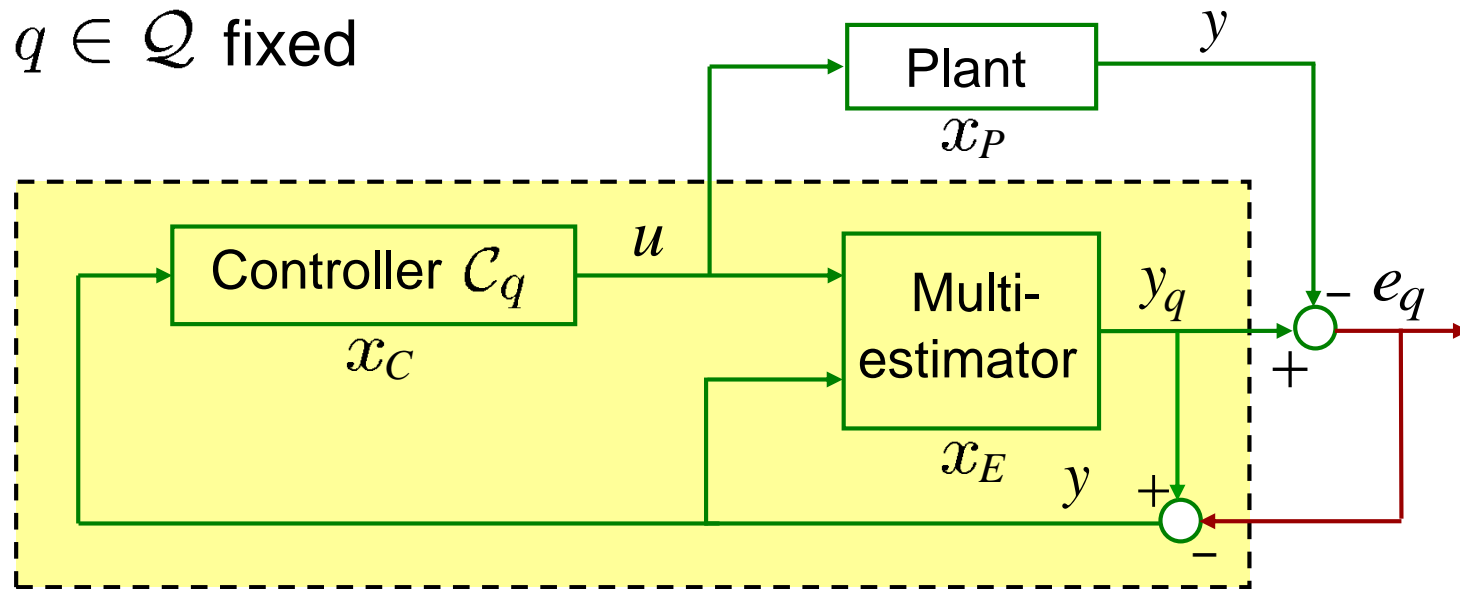
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Nonlinear: same result holds if stability and detectability are interpreted in the **integral-ISS/OSS** sense:

$$|x(t)| \leq \beta(|x(0)|, t) + \int_0^t \gamma(|v(\tau)|) d\tau$$

CANDIDATE CONTROLLERS

$q \in \mathcal{Q}$ fixed

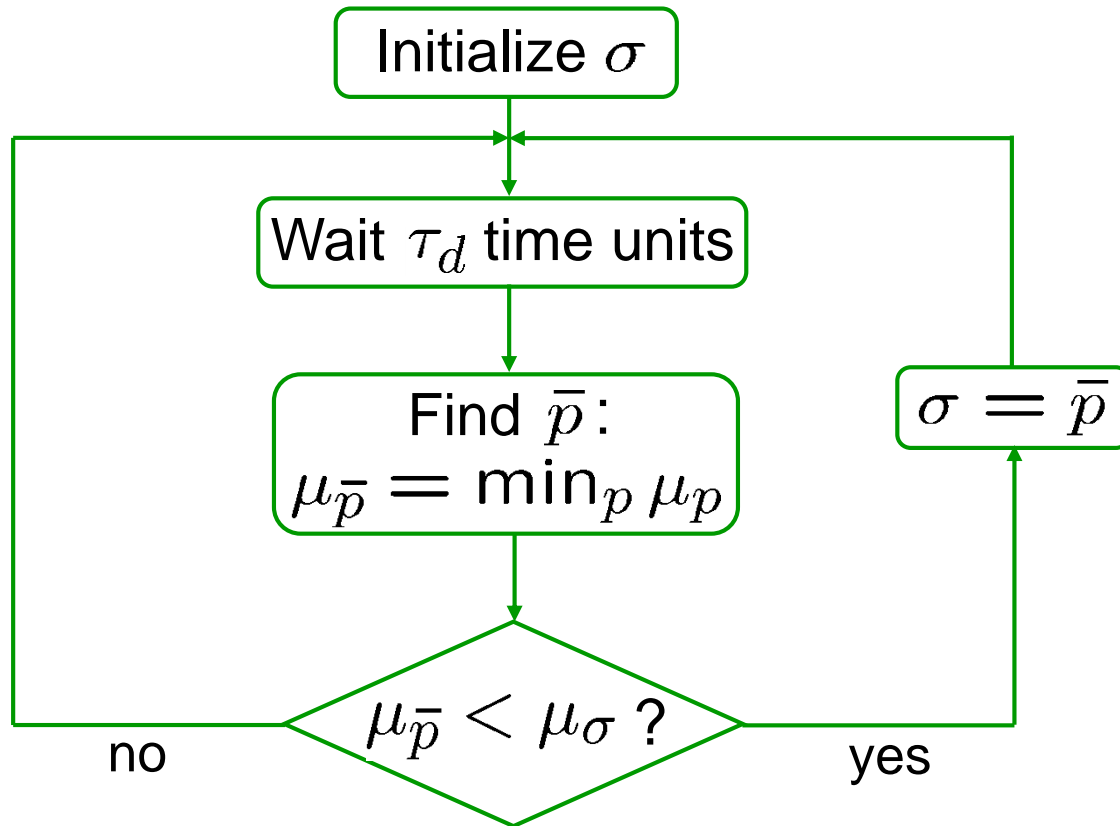


- Linear:** overall system is detectable w.r.t. e_q if
- system inside the box is stable
 - plant is detectable

For minimum-phase plants, it is enough to ask that the system inside the box be **output**-stabilized

SWITCHING LOGIC: DWELL-TIME

$\mu_p, p \in \mathcal{P}$ – monitoring signals $\tau_d > 0$ – dwell time



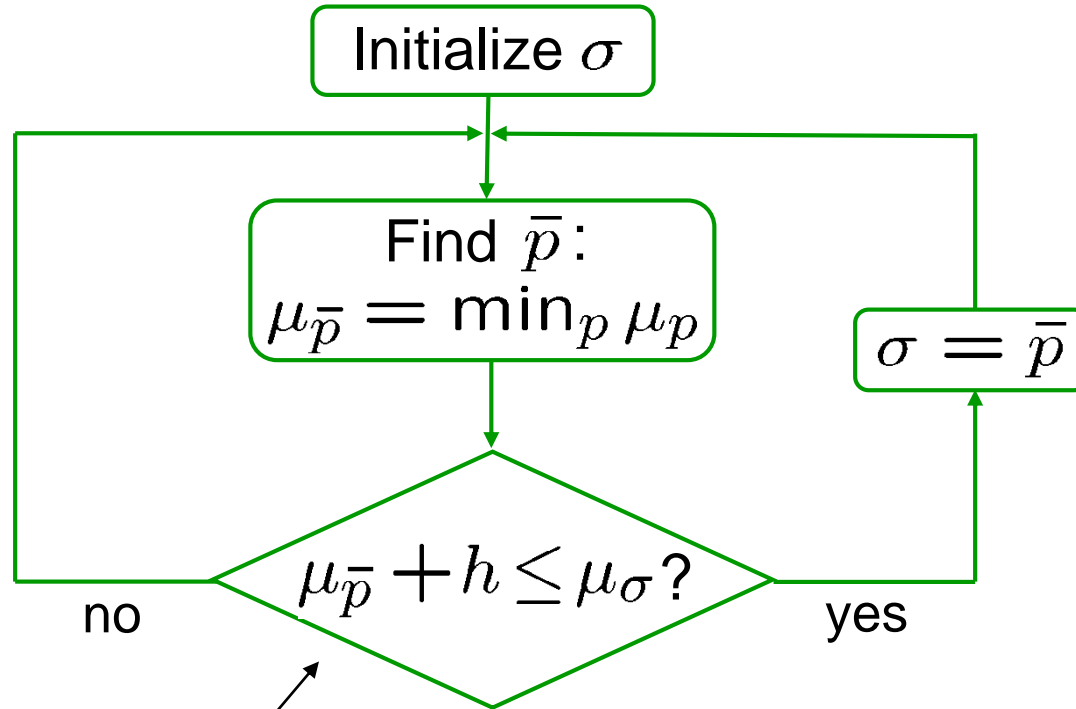
Detectability is preserved if τ_d is large enough ✓

Obtaining a bound on e_{σ} in terms of e_{p^*} is harder

Not suitable for nonlinear systems (finite escape)

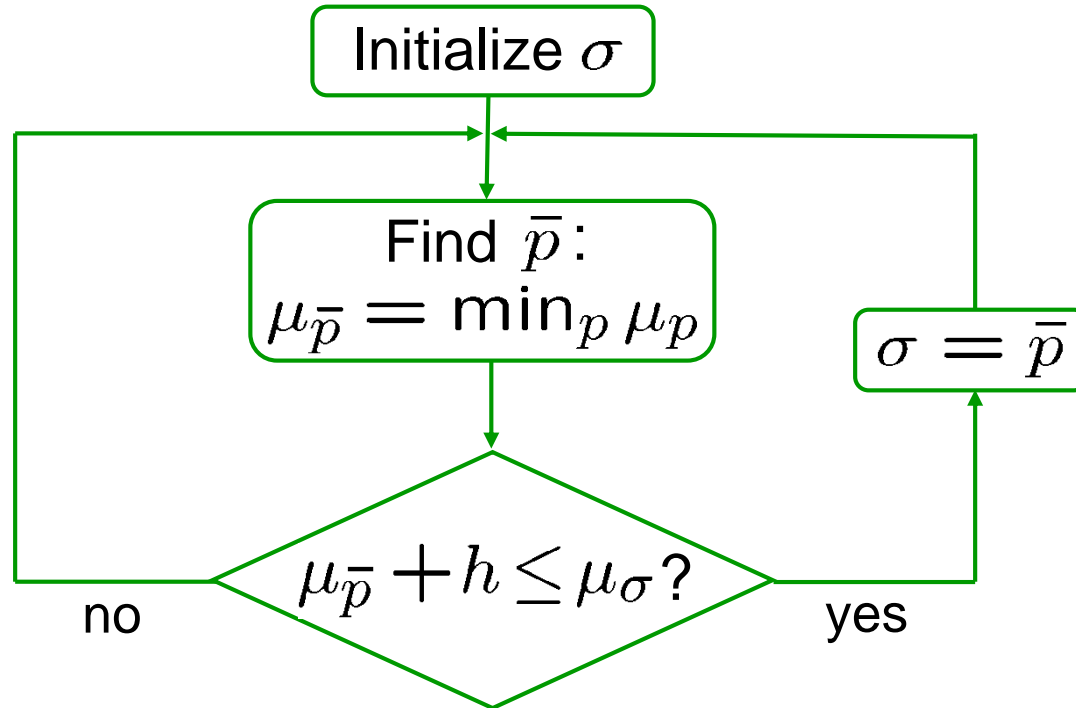
SWITCHING LOGIC: HYSTERESIS

$\mu_p, p \in \mathcal{P}$ – monitoring signals $h > 0$ – hysteresis constant



or $(1 + h)\mu_{\bar{p}} \leq \mu_{\sigma}$
(scale-independent)

SWITCHING LOGIC: HYSTERESIS



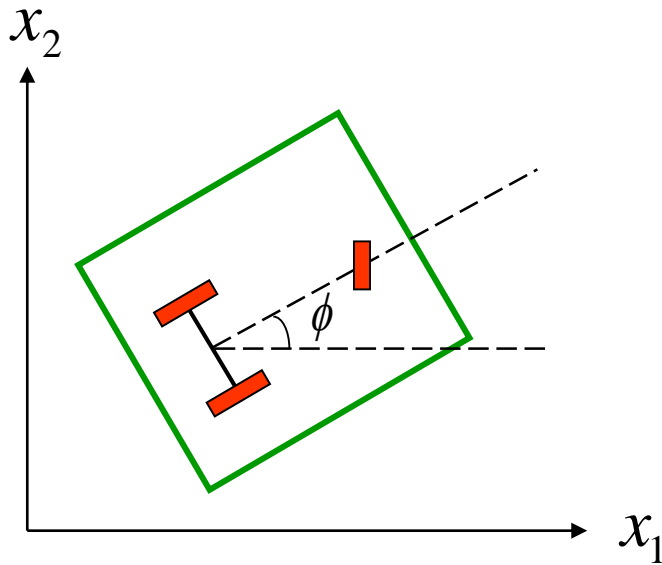
\mathcal{P} finite, $\mu_p \uparrow$, μ_p^* bounded \Rightarrow switching **stops in finite time**

This applies to $\delta, n, d = 0$, $e_p^* \rightarrow 0$ exp fast, $\mu_p = \int |e_p|^2$

Linear, $\delta = 0$, n, d bounded \Rightarrow **average dwell time** $\tau_a(h)$

$$\int |e_\sigma|^2 \leq |\mathcal{P}|(1+h) \int |e_p|^2$$

TOY EXAMPLE: PARKING PROBLEM



$$\dot{x}_1 = p_1 w_1 \cos \phi$$

$$\dot{x}_2 = p_1 w_1 \sin \phi$$

$$\dot{\phi} = p_2 w_2$$

Unknown parameters p_1, p_2 correspond to the radius of rear wheels and distance between them